

Numerical evaluation of a two-body point absorber wave energy converter with a tuned inerter

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Abstract

Ocean wave energy has been considered as one of the prospective renewable energy sources. To increase the amount of energy captured from ocean waves more, this article proposes a two-body point absorber wave energy converter (WEC) with a tuned inerter. The tuned inerter mechanism employed here consists of a spring, a linear damping element, and an inerter. This mechanism was originally proposed in the field of civil engineering as a structural control device which can absorb energy from vibrating structures effectively by taking advantage of the resonance effect of the inerter. In addition to the tuned inerter mechanism using a generator as the linear damping element, the current of the generator for the power take-off system is controlled based on the algorithms proposed in literature to achieve further improvement of the power generation performance. In this research, a detailed analytical model of the proposed two-body point absorber WEC is introduced and the equation of motion and the state-space representation are developed. Then the power generation performance for the JONSWAP spectrum is assessed and the effectiveness of the tuned inerter mechanism for two-body point absorber WECs is shown through numerical simulation studies.

Keywords: wave energy converter, two-body point absorber, tuned inerter, renewable energy, energy harvesting

1. Introduction

Ocean has been expected to be a promising renewable energy source since more than 70% of the Earth's surface is covered with oceans. However, compared with other renewable energy sources such as wind and solar energy,

ocean energy conversion technology has not yet shown a strong presence in the renewable energy market. Since the concept of wave energy converter (WEC) was introduced by a former Japanese naval commander, Yoshio Masuda (1925-2009) [1], considerable effort has been devoted to develop a variety of WECs [2, 3, 4, 5] to exploit wave power in the ocean effectively. And various types of WECs proposed so far includes oscillating water columns [6, 7], oscillating bodies [8, 9, 10], and overtopping devices [11, 12, 13]. Among these devices, point absorbers consisting floating bodies are categorized as the oscillating body WEC and more expectations have been placed on this type of WEC because of its availability in deep offshore regions and its extensibility by arraying many buoys.

To improve the power generation performance of a conventional single-body point absorber WEC, which has one floating buoy, the authors employed a tuned inerter mechanism [14, 15]. Originally, the tuned inerter mechanism was proposed by [16] as a structural control device to absorb vibration energy effectively from vibrating civil structures such as buildings induced by seismic disturbances and to mitigate damage. This mechanism consists of a tuning spring, a linear damping element, and an inerter [17]. The inerter is an element to produce a force proportional to the difference between the accelerations of both ends and realized by devices such as a ball screw and a rack and pinion. In the tuned inerter mechanism, the damping element is installed in parallel with the inerter and these two elements are connected to the spring in series. Thus, once the spring stiffness is tuned so that the inerter resonates with the dominant frequency of the input vibration to the device, the deformation of the damping part is increased and the vibration energy is dissipated more effectively.

The authors employed a generator or a motor as the linear damping element in the tuned inerter mechanism and showed that the mechanism enhances the ability to extract energy from vibrating structures at a low frequency of less than 10 Hz [18, 19]. In addition, the authors proposed a single-body point absorber WEC with a tuned inerter and the efficacy of the present device was shown through numerical simulation studies [14] and wave flume testing using a small-scale prototype model [15].

While, generally, in the research community of point absorber WECs, a two-body point absorber WEC [20, 21, 22, 23] consisting a floating buoy and a submerged body is considered a preferable device from the view point of cost and deployment. This is because it does not need to connect to the ocean floor directly. Additionally, the two-body type has the potential to

improve the power generation capability more than the single-body type by designing the two bodies to achieve a greater relative velocity between the two bodies.

The primary objective of this paper is to propose a two-body point absorber WEC with a tuned inerter. A detailed analytical model including the coupled force between the two bodies for the proposed WEC and the drag force is introduced and the equation of motion and the state-space representation are developed. Then the energy harvesting performance for the JONSWAP spectrum is assessed by comparing with a typical two-body point absorber WEC without the proposed mechanism. Secondly, the effectiveness of the current control of the generator for the power take-off (PTO) system of the proposed WEC is investigated. As control algorithms, two controllers proposed in [24], i.e., static admittance (SA) control and performance guaranteed (PG) control, are applied to capture more energy. The obtained results of the numerical studies using WATMIT [25] are shown and conclusions gained from this research follow.

It should be noted that we make use of the short-hand $\mathbf{G} \sim \left[\begin{array}{c|c} \mathbf{A} & \mathbf{B} \\ \hline \mathbf{C} & \mathbf{D} \end{array} \right]$ to imply $\mathbf{G}(s) = \mathbf{C}[s\mathbf{I} - \mathbf{A}]^{-1}\mathbf{B} + \mathbf{D}$ and $\hat{f}(\omega)$ denotes the Fourier transform of a function $f(t)$ in this article. Also, note that j is the imaginary unit such that $j^2 = -1$ and that the expected value is denoted by $\mathcal{E}\{\cdot\}$.

2. Modeling

To implement numerical studies, the analytical model of the proposed WEC is developed here as well as the models of the hydrodynamic forces, drag forces, and the JONSWAP spectrum. For simplicity, we consider the heave direction only because this motion becomes dominant for the power extraction of wave energy. Additionally, the generated power is defined in this section. Note that the floating buoy and the submerged body are indicated by 1st and 2nd bodies, respectively, in this article.

2.1. Conventional two-body point absorber WEC

For comparison, a conventional two-body point absorber WEC consisting a floating buoy (1st body) and a submerged body (2nd body) shown schematically in Fig. 1 (a) is reviewed briefly first. A PTO system including a generator is placed between these two bodies. In this research, it is assumed that the floating buoy has a circular cylinder shape with a diameter D_1 and

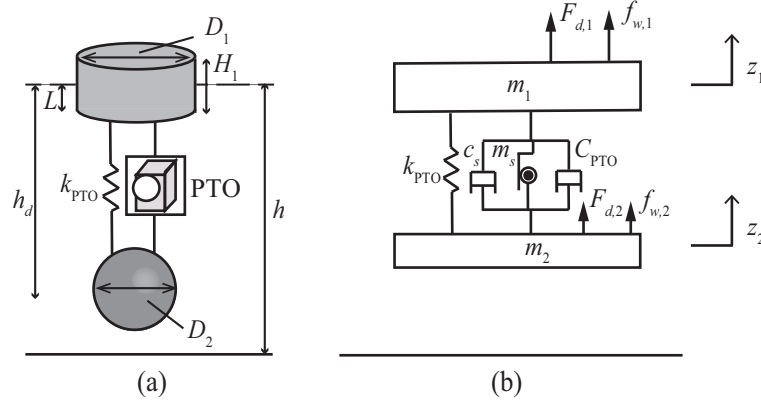


Figure 1: Conventional two-body point absorber WEC: (a) Schematic illustration, (b) Model.

the submerged body is a sphere with a diameter D_2 . Also, the distance between these bodies is h_d at the static equilibrium position and these bodies are connected by a spring whose stiffness is k_{PTO} .

The model of the conventional type is illustrated in Fig. 1 (b). Let z_k , ($k = 1, 2$) be the displacement of the k th body and z_s be the rotational displacement of the generator. Then we have

$$z_s = z_1 - z_2 \quad (1)$$

Hence the equation of motion of the floating buoy would be

$$m_1 \ddot{z}_1 + m_s (\ddot{z}_1 - \ddot{z}_2) + (c_s + C_{\text{PTO}}) (\dot{z}_1 - \dot{z}_2) + k_{\text{PTO}} (z_1 - z_2) = F_{d,1} + f_{w,1} \quad (2)$$

where m_1 is the mass of the floating buoy, m_s is the inertance caused by the generator itself, c_s is the unwanted inherent mechanical damping coefficient caused in the PTO system, C_{PTO} is the damping coefficient of the generator. Also, $f_{w,1}$ and $F_{d,1}$ are the hydrodynamic force and the drag force acting on the floating buoy, respectively.

While, the equation of motion of the submerged body is developed as

$$m_2 \ddot{z}_2 - m_s (\ddot{z}_1 - \ddot{z}_2) - (c_s + C_{\text{PTO}}) (\dot{z}_1 - \dot{z}_2) - k_{\text{PTO}} (z_1 - z_2) = F_{d,2} + f_{w,2} \quad (3)$$

where m_2 is the mass of the submerged body and $f_{w,2}$ and $F_{d,2}$ are the hydrodynamic force and the drag force on the submerged body, respectively, similarly to the floating buoy.

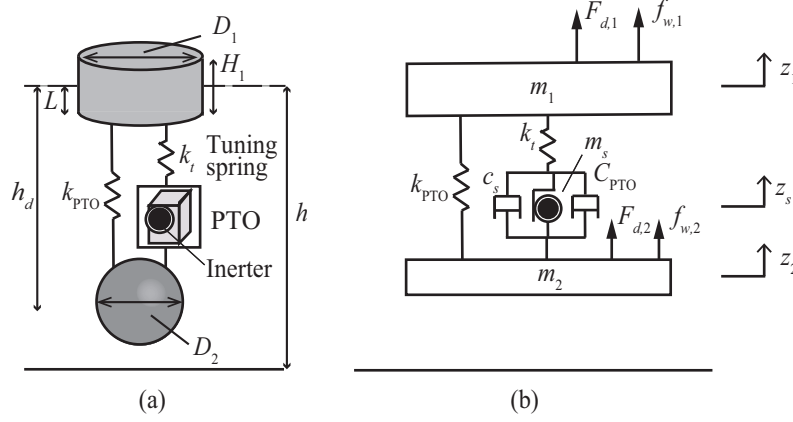


Figure 2: Two-body point absorber WEC with a tuned inerter: (a) Schematic illustration, (b) Model.

2.2. Two-body point absorber WEC with a tuned inerter

Next, the proposed two-body point absorber WEC with a tuned inerter shown in Fig. 2 (a) is considered. As can be seen, unlike the conventional two-body type, a tuning spring whose stiffness is k_t is added between the floating buoy and the PTO system. Also, an additional rotational mass such as a flywheel producing sufficiently large inertance m_s is mounted intentionally on the generator shaft.

The model of the present device is shown in Fig. 2 (b) and the equations of motion of the device are derived as follows. In contrast to the conventional type, Eq. (1) is not satisfied because the proposed system becomes a three-degree-of-freedom system due to the tuning spring. Then, for this model, the equations of motion of the floating buoy and the submerged body are given by

$$m_1 \ddot{z}_1 + k_{\text{PTO}}(z_1 - z_2) + k_t(z_1 - z_2 - z_s) = F_{d,1} + f_{w,1} \quad (4)$$

$$m_2 \ddot{z}_2 - k_{\text{PTO}}(z_1 - z_2) - k_t(z_1 - z_2 - z_s) = F_{d,2} + f_{w,2} \quad (5)$$

respectively. And considering the fact that the force of the tuning spring equals the force of the PTO system, the equation of motion of the inerter is derived as

$$m_s \ddot{z}_s + (c_s + C_{\text{PTO}}) \dot{z}_s = k_t(z_1 - z_2 - z_s) \quad (6)$$

110 2.3. Hydrodynamic force

111 The hydrodynamic force $f_{w,k}$ acting on the k th body is described based
112 on the linear potential wave theory by

$$f_{w,k} = f_{a,k} + f_{b,k} + f_{c,k} \quad (7)$$

113 where $f_{a,k}$ is the excitation force, $f_{b,k}$ is the hydrodynamic forces due to
114 buoyancy, and $f_{c,k}$ is the radiation force.

115 The relationship between the excitation force $f_{a,k}$ and the amplitude of
116 the incident wave $a(t)$ is given in the frequency domain using a transfer
117 function $F_{a,k}(\omega)$ as

$$\hat{f}_{a,k}(\omega) = F_{a,k}(\omega)\hat{a}(\omega) \quad (8)$$

118 The hydrostatic force $f_{b,1}$ on the cylindrical floating buoy becomes a linear
119 function of z_1 given as

$$f_{b,1} = -K_w z_1, \quad K_w = \rho g \pi \left(\frac{D_1}{2} \right)^2 \quad (9)$$

120 where g is gravitational acceleration and ρ is the sea water density. While
121 the hydrostatic force of the submerged body is constant, thus $f_{b,2}$ can be set
122 as

$$f_{b,2} = 0 \quad (10)$$

123 at the equilibrium position in the equation of motion.

124 Next, define $l = 1, 2, (l \neq k)$, then the radiation force $f_{c,k}$ on the k th body
125 including the coupled force affected by the l th body is given by

$$\hat{f}_{c,k} = -(j\omega A_{kk}(\omega) + B_{kk}(\omega))\hat{z}_k - (j\omega A_{kl}(\omega) + B_{kl}(\omega))\hat{z}_l \quad (11)$$

126 where A_{kk} and B_{kk} are the added mass and the radiation damping of the k th
127 body, and A_{kl} and B_{kl} represent the coupled added mass and the coupled
128 radiation damping from the l th body to the k th body.

129 2.4. Drag force

130 The drag force $F_{d,k}$ acting on the k th body is modeled according to the
131 nonlinear Morison equation given by [26]

$$F_{d,k} = -\frac{1}{2}\rho S_k C_{d,k} |\dot{z}_k| \dot{z}_k \quad (12)$$

132 where S_k is the characteristic area and $C_{d,k}$ is the dimensionless drag coefficient.
 133 As stated before, in this research, the shapes of the floating buoy and
 134 the submerged body are assumed to be a cylinder and a sphere, respectively,
 135 thus we have

$$S_1 = \frac{\pi D_1^2}{4}, \quad S_2 = \frac{\pi D_2^2}{4} \quad (13)$$

136 However, it would be cumbersome to deal with nonlinear equations in
 137 the frequency domain, Eq. (12) is linearized under the condition of irregular
 138 wave as [27, 26]

$$F_{d,k} = -\frac{1}{2}\rho S_k C_{d,k} \sqrt{\frac{8}{\pi}} \sigma_{\dot{z}_k} \dot{z}_k \quad (14)$$

139 where $\sigma_{\dot{z}_k}$ is the standard deviation of $\dot{z}_k$. While, in general, the drag force
 140 of the floating buoy is negligible compared to the hydrodynamic force, thus
 141 we assume that

$$F_{d,1} = 0 \quad (15)$$

142 From, Eq. (14), the linearized drag force on the submerged body is given
 143 with the viscous damping coefficient $c_{v,2}$ by

$$F_{d,2} = -c_{v,2} \dot{z}_2, \quad c_{v,2} = \frac{1}{2}\rho S_2 C_{d,2} \sqrt{\frac{8}{\pi}} \sigma_{\dot{z}_2} \quad (16)$$

144 Hereafter, in this article, $\sigma_{\dot{z}}$ is used to represent the standard deviation of $\dot{z}_2$
 145 for simplicity.

146 2.5. Stochastic sea state model

147 In simple theoretical models of WECs, it is typical to assume the incident
 148 waves to be regular. For a more realistic model, irregular waves are used with
 149 time-domain analysis which requires much more computing time [4]. An
 150 alternative method with less computation for modeling true sea states is the
 151 stochastic modeling. We assume the wave amplitude $a(t)$ to be a stationary
 152 stochastic process with spectral density $S_a(\omega)$ and variance given by

$$\sigma_a^2 = \frac{1}{2\pi} \int_{-\infty}^{\infty} S_a(\omega) d\omega \quad (17)$$

153 Characterizing $S_a(\omega)$ by the JONSWAP spectrum [28] with its mean wave
 154 period T_1 , significant wave height H_s , and peak enhancement factor γ derives

155 the following

$$S_a(\omega) = 310\pi \frac{H_s^2}{T_1^4 \omega^5} \exp \left[\frac{-944}{T_1^4 \omega^4} \right] \gamma^W \quad (18)$$

156 where

$$W = \exp \left[- \left(\frac{0.191\omega T_1 - 1}{\sqrt{2}\sigma} \right)^2 \right] \quad (19)$$

157 and

$$\sigma = \begin{cases} 0.07 & : \omega T_1 \leq 5.24 \\ 0.09 & : \omega T_1 > 5.24 \end{cases} \quad (20)$$

158 The JONSWAP spectrum can also be represented by the peak period T_p
159 using the well-known relationship

$$T_1 = 0.834T_p \quad (21)$$

160 2.6. Power take-off system

161 In this study, the generator is assumed to be a three-phase permanent
162 magnet synchronous machine (PMSM). However, the three phase voltage
163 and current vectors can be transformed to "quadrature components". More
164 details on this transformation can be found in [24, 29]. Then, assuming an
165 ideal generator with linear behavior and minimal core loss results in linearity
166 between the back-EMF e and the velocity coupled with the generator $\dot{z}_s$.
167 Therefore, the equation relating to e and $\dot{z}_s$ is given as

$$e = K_e \dot{z}_s \quad (22)$$

168 where K_e is a constant associated with the back-EMF of the generator. By
169 reciprocity, the electromagnetic force and generator current i has the follow-
170 ing linear relationship

$$C_{PTO} \dot{z}_s = -K_e i \quad (23)$$

171 In the case of single-directional converter used in [24], the input current
172 to the generator i can be expressed as

$$i = -Y e \quad (24)$$

173 where Y is the admittance of the generator restricted in ideal conditions by
174 [24]

$$Y \in [0, 1/R] \quad (25)$$

175 and R is the internal or coil resistance of the generator. Applying Eq. (24)
 176 to Eq. (23) with Eq. (22) yields

$$C_{\text{PTO}} = Y K_e^2 \quad (26)$$

177 which expresses how the generator damping C_{PTO} is controlled by the ad-
 178 mittance Y .

179 The total power generation is defined as the extracted power minus the
 180 electrical loss [10]. In this paper, we assume that the current-dependent loss
 181 is resistive, i.e., Ri^2 , then we have the power generation as

$$P_g = -ei - Ri^2 \quad (27)$$

182 3. State-space representation

183 In this section, to assess the power generation by the current controllers
 184 under stochastic sea states, a state-space form of the proposed WEC aug-
 185 mented with the JONSWAP spectrum is developed. Also, the controllers for
 186 the current of the generator for power generation are reviewed briefly.

187 3.1. JONSWAP spectrum

188 First, a state-space model of the wave amplitude is derived. We find a
 189 finite-dimensional noise filter

$$F_w \sim \begin{bmatrix} \mathbf{A}_w & \mathbf{B}_w \\ \mathbf{C}_w & \mathbf{0} \end{bmatrix} \quad (28)$$

190 such that its power spectrum is close to the JONSWAP spectrum, i.e.,
 191 $S_a(\omega) = |F_w(\omega)|^2$, for a unit intensity white noise input $w(t)$. Then we
 192 have

$$\dot{\mathbf{x}}_w(t) = \mathbf{A}_w \mathbf{x}_w(t) + \mathbf{B}w(t) \quad (29)$$

$$a(t) = \mathbf{C}_w \mathbf{x}_w(t) \quad (30)$$

193 According to the simplified procedure advocated by Spanos [30], F_w can be
 194 approximated by a forth-order controllable canonical form of

$$\mathbf{A}_w = \begin{bmatrix} 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ a_1 & a_2 & a_3 & a_4 \end{bmatrix}, \quad \mathbf{B}_w = \begin{bmatrix} 0 \\ 0 \\ 0 \\ 1 \end{bmatrix}, \quad \mathbf{C}_w = [0 \quad 0 \quad c_3 \quad 0] \quad (31)$$

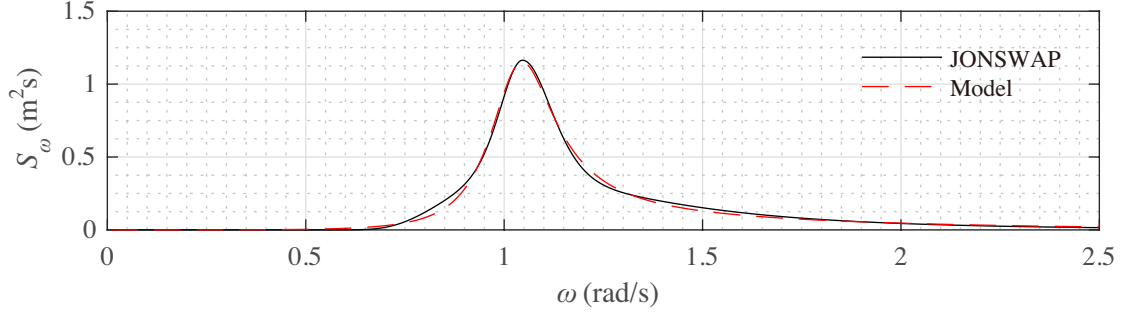


Figure 3: JONSWAP spectrum with $T_p = 6$ s, $H_s = 1$ m, $\gamma = 3.3$

where the filter parameters a_1 , a_2 , a_3 , a_4 , and c_3 are chosen to minimize the mean-square error $\int_{-\infty}^{\infty} (S_a(\omega) - |F_w(\omega)|^2)^2 d\omega$, while constraining a_1 through a_4 so that the system poles are in the open left half plane.

For example, Fig. 3 shows a JONSWAP spectrum for $T_p = 6$ s, $H_s = 1$ m, $\gamma = 3.3$ and its fourth-order finite-dimensional approximate system. We can confirm in the figure that the fourth-order F_w estimates the JONSWAP spectrum very well. It should be noted that $H_s = 1$ m, $\gamma = 3.3$ are fixed in the numerical simulation studies in this paper.

3.2. Two-body point absorber WEC with a tuned inerter

Next, state-space representation for the proposed device is developed here. As state-space representation for the conventional two-body WEC can be developed in a similar way, its derivation is omitted.

Substitute the equations for the hydrodynamic and drag forces into Eqs. (4) and (5), then taking Fourier transform gives

$$\begin{aligned} & \{-\omega^2(m_1 + A_{11}(\omega)) + j\omega(c_{v1} + B_{11}(\omega)) + (K_w + k_{PTO} + k_t)\} \hat{z}_1 \\ & + \{-\omega^2 A_{12}(\omega) + j\omega B_{12}(\omega) - (k_{PTO} + k_t)\} \hat{z}_2 \\ & = F_{a,1}(\omega) \hat{a} + k_t \hat{z}_s \end{aligned} \quad (32)$$

and

$$\begin{aligned} & \{-\omega^2 A_{21}(\omega) + j\omega B_{21}(\omega) - (k_{PTO} + k_t)\} \hat{z}_1 \\ & + \{-\omega^2(m_2 + A_{22}(\omega)) + j\omega(c_{v2} + B_{22}(\omega)) + (k_{PTO} + k_t)\} \hat{z}_2 \\ & = F_{a,2}(\omega) \hat{a} - k_t \hat{z}_s \end{aligned} \quad (33)$$

210 respectively. Hence, Eqs. (32) and (33) can be combined and written in
 211 matrix form as

$$\begin{bmatrix} Z_{11} & Z_{12} \\ Z_{21} & Z_{22} \end{bmatrix} \begin{bmatrix} \hat{z}_1 \\ \hat{z}_2 \end{bmatrix} = \begin{bmatrix} F_{a,1} \\ F_{a,2} \end{bmatrix} \hat{a} + \begin{bmatrix} k_t \\ -k_t \end{bmatrix} \hat{z}_s \quad (34)$$

212 where

$$\begin{aligned} Z_{11}(\omega) &= -\omega^2(m_1 + A_{11}(\omega)) + j\omega(B_{11}(\omega)) + (K_w + k_{\text{PTO}} + k_t) \\ Z_{12}(\omega) &= -\omega^2 A_{12}(\omega) + j\omega B_{12}(\omega) - (k_{\text{PTO}} + k_t) \\ Z_{21}(\omega) &= -\omega^2 A_{21}(\omega) + j\omega B_{21}(\omega) - (k_{\text{PTO}} + k_t) \\ Z_{22}(\omega) &= -\omega^2(m_2 + A_{22}(\omega)) + j\omega(B_{22}(\omega)) + (k_{\text{PTO}} + k_t) \end{aligned} \quad (35)$$

213 Solving Eq. (34) for $[\hat{z}_1 \ \hat{z}_2]^T$ yields the expressions of form

$$\hat{z}_1 = G_{a,1}(\omega)\hat{a} + G_{z,1}(\omega)\hat{z}_s \quad (36)$$

$$\hat{z}_2 = G_{a,2}(\omega)\hat{a} + G_{z,2}(\omega)\hat{z}_s \quad (37)$$

214 When $G_{a,1}$, $G_{z,1}$, $G_{a,2}$, and $G_{z,2}$ are approximated by finite-dimensional
 215 systems, we have representations as

$$G_{a,1} \sim \begin{bmatrix} \mathbf{A}_{a,1} & \mathbf{B}_{a,1} \\ \mathbf{C}_{a,1} & \mathbf{0} \end{bmatrix}, \quad G_{z,1} \sim \begin{bmatrix} \mathbf{A}_{z,1} & \mathbf{B}_{z,1} \\ \mathbf{C}_{z,1} & \mathbf{0} \end{bmatrix} \quad (38)$$

216

$$G_{a,2} \sim \begin{bmatrix} \mathbf{A}_{a,2} & \mathbf{B}_{a,2} \\ \mathbf{C}_{a,2} & \mathbf{0} \end{bmatrix}, \quad G_{z,2} \sim \begin{bmatrix} \mathbf{A}_{z,2} & \mathbf{B}_{z,2} \\ \mathbf{C}_{z,2} & \mathbf{0} \end{bmatrix} \quad (39)$$

217 It should be noted that the function $F_{a,k}$ in Eq. (8) is non-causal which will
 218 be problematic when approximating $G_{a,1}$ and $G_{a,2}$ by a finite-dimensional
 219 state-space. Therefore the technique of spatial delay proposed by Falnes [31]
 220 is used, defining $a(t)$ as the wave amplitude at a distance of d in front of the
 221 buoy.

222 Once Eqs. (38) and (39) are obtained by system identification techniques,
 223 the identified systems Eqs. (36) and (37) are represented in the time domain
 224 as

$$\dot{\mathbf{x}}_1(t) = \mathbf{A}_1 \mathbf{x}_1(t) + \mathbf{G}_1 a(t) + \mathbf{E}_1 z_s(t) \quad (40)$$

$$z_1(t) = \mathbf{C}_1 \mathbf{x}_1(t) \quad (41)$$

where $\mathbf{A}_1$, $\mathbf{G}_1$, $\mathbf{E}_1$, and $\mathbf{C}_1$ are expressed by $\mathbf{A}_{a,1}$, $\mathbf{B}_{a,1}$, $\mathbf{C}_{a,1}$, $\mathbf{A}_{z,1}$, $\mathbf{B}_{z,1}$, and $\mathbf{C}_{z,1}$. and

$$\dot{\mathbf{x}}_2(t) = \mathbf{A}_2 \mathbf{x}_2(t) + \mathbf{G}_2 a(t) + \mathbf{E}_2 z_s(t) \quad (42)$$

$$z_2(t) = \mathbf{C}_2 \mathbf{x}_2(t) \quad (43)$$

where $\mathbf{A}_2$, $\mathbf{G}_2$, $\mathbf{E}_2$, and $\mathbf{C}_2$ are expressed by $\mathbf{A}_{a,2}$, $\mathbf{B}_{a,2}$, $\mathbf{C}_{a,2}$, $\mathbf{A}_{z,2}$, $\mathbf{B}_{z,2}$, and $\mathbf{C}_{z,2}$ as well.

Also, considering $z_1 - z_2$ as an input, we have a state-space representation about the tuned inerter part where the output is the velocity of the generator as

$$\dot{\mathbf{x}}_s(t) = \mathbf{A}_s \mathbf{x}_s(t) + \mathbf{B}_s i(t) + \mathbf{E}_s (z_1(t) - z_2(t)) \quad (44)$$

$$\dot{z}_s(t) = \mathbf{C}_s \mathbf{x}_s(t) \quad (45)$$

where the state vector is defined as $\mathbf{x}_s = [z_s \ \dot{z}_s]^T$ and

$$\mathbf{A}_s = \begin{bmatrix} 0 & 1 \\ -\frac{k_t}{m_s} & -\frac{c_s}{m_s} \end{bmatrix}, \quad \mathbf{B}_s = \begin{bmatrix} 0 \\ \frac{c_c}{m_s} \end{bmatrix}, \quad \mathbf{E}_s = \begin{bmatrix} 0 \\ \frac{k_t}{m_s} \end{bmatrix}, \quad \mathbf{C}_s = [0 \ 1] \quad (46)$$

Define the state vector as $\mathbf{x}_h = [\mathbf{x}_1^T \ \mathbf{x}_2^T \ \mathbf{x}_s^T]^T$. Then we have a state-space representation in which the inputs are the current i and the wave height a and the output is the voltage e from Eq. (22) and Eqs. (40) through (46) as follows:

$$\dot{\mathbf{x}}_h(t) = \mathbf{A}_h \mathbf{x}_h(t) + \mathbf{B}_h i(t) + \mathbf{G}_h a(t) \quad (47)$$

$$e(t) = \mathbf{C}_h \mathbf{x}_h(t) \quad (48)$$

where $\mathbf{A}_h$, $\mathbf{B}_h$, $\mathbf{G}_h$, and $\mathbf{C}_h$ can be composed of $\mathbf{A}_1$, $\mathbf{A}_2$, $\mathbf{B}_1$, $\mathbf{B}_2$, $\mathbf{G}_1$, $\mathbf{G}_2$, $\mathbf{C}_1$, $\mathbf{C}_2$, $\mathbf{A}_s$, $\mathbf{B}_s$, $\mathbf{E}_s$, and $\mathbf{C}_s$.

3.3. Augmented system

Finally, combining Eqs. (47) and (48) with the stochastic sea state model given by Eqs. (29) and (30) gives the augmented system where the external disturbance input is white noise $w(t)$ expressed as

$$\dot{\mathbf{x}}(t) = \mathbf{A} \mathbf{x}(t) + \mathbf{B} i(t) + \mathbf{G} w(t) \quad (49)$$

$$e(t) = \mathbf{C} \mathbf{x}(t) \quad (50)$$

where

$$\mathbf{x} = \begin{bmatrix} \mathbf{x}_h \\ \mathbf{x}_w \end{bmatrix}, \quad \mathbf{A} = \begin{bmatrix} \mathbf{A}_h & \mathbf{G}_h \mathbf{C}_w \\ \mathbf{0} & \mathbf{A}_w \end{bmatrix}, \quad \mathbf{B} = \begin{bmatrix} \mathbf{B}_h \\ \mathbf{0} \end{bmatrix}, \quad \mathbf{G} = \begin{bmatrix} \mathbf{0} \\ \mathbf{B}_w \end{bmatrix}, \quad \mathbf{C} = [\mathbf{C}_h \ \mathbf{0}] \quad (51)$$

244 3.4. Controller

245 To improve the power generation performance and to examine the effec-
 246 tiveness of the current control on the proposed device, the current of the
 247 generator i is controlled by the two control laws introduced in [24] which are
 248 reviewed briefly here. Before we go any further, it should be noted that the
 249 average of the generated power defined as Eq. (27) can be written as

$$\bar{P}_g = -\varepsilon \left\{ \begin{bmatrix} \mathbf{x} \\ i \end{bmatrix}^T \begin{bmatrix} \mathbf{0} & \frac{1}{2}\mathbf{C}^T \\ \frac{1}{2}\mathbf{C} & R \end{bmatrix} \begin{bmatrix} \mathbf{x} \\ i \end{bmatrix} \right\} \quad (52)$$

250 using Eq. (50).

251 3.4.1. Static admittance control

252 For the SA control, a constant feedback gain Y_c restricted by Eq. (25) is
 253 adopted so that the value defined by Eq. (52) is maximized. The method to
 254 search for such a value is reviewed here.

255 From Eqs. (24) and (50), the input current to the generator is expressed
 256 as a function of the state variable $\mathbf{x}$, i.e.,

$$i(t) = -Y_c \mathbf{C} \mathbf{x}(t) \quad (53)$$

257 Substituting Eq. (53) into Eq. (49) yields the closed-loop dynamics having
 258 the form

$$\dot{\mathbf{x}}(t) = (\mathbf{A} - Y_c \mathbf{B} \mathbf{C}) \mathbf{x}(t) + \mathbf{G} w(t) \quad (54)$$

259 Let the average power by the SA control be $\bar{P}_g^{\text{SA}}$. Then for any time-invariant
 260 Y_c satisfying Eq. (25), it is a standard result that the power generation
 261 objective can be written as [32]

$$\bar{P}_g^{\text{SA}} = -\text{tr}[\mathbf{G}^T \mathbf{S} \mathbf{G}] \quad (55)$$

262 where $\mathbf{S} = \mathbf{S}^T < 0$ is the solution to the Lyapunov equation

$$(\mathbf{A} - Y_c \mathbf{B} \mathbf{C})^T \mathbf{S} + \mathbf{S} (\mathbf{A} - Y_c \mathbf{B} \mathbf{C}) + \mathbf{C}^T (-Y_c + Y_c^2 R) \mathbf{C} = \mathbf{0} \quad (56)$$

263 Y_c must be less than or equal to $1/R$, so the last term on the left-hand side
 264 of Eq. (56) is negative-semidefinite for all Y_c . Thus, since $\mathbf{A} - Y_c \mathbf{B} \mathbf{C}$ is
 265 asymptotically stable, the definiteness of $\mathbf{S}$ is assured by Lyapunov's second
 266 theorem [33]. Then the optimal value for Y_c is chosen so that Eq. (55) is
 267 maximized.

268 3.4.2. Performance guaranteed control

269 For comparison, the efficacy of time-varying gain Y based on the PG
 270 control algorithm proposed in the literature is examined. This algorithm is
 271 operated with a single-directional converter and the admittance Y becomes
 272 a function of time varying within the range of Eq. (25) so that the generated
 273 average power $\bar{P}_g^{\text{PG}}$ must be larger than $\bar{P}_g^{\text{SA}}$, i.e.,

$$\bar{P}_g^{\text{PG}} \geq \bar{P}_g^{\text{SA}} \quad (57)$$

274 In this algorithm, the admittance is controlled by

$$Y(t) = \text{sat}_{[0, 1/R]} \left\{ \frac{\mathbf{K}\mathbf{x}}{e} \right\} \quad (58)$$

275 where

$$\mathbf{K} = -\frac{1}{R} \left(\mathbf{B}^T \mathbf{S} + \frac{1}{2} \mathbf{C} \right) \quad (59)$$

276 Note that Y_c is the constant value for the SA control. In this case, the current
 277 i is expressed by

$$i(t) = \begin{cases} i_u(t) & : i_u e + i_u^2 R \leq 0 \\ 0 & : i_u e + i_u^2 R > 0 \text{ and } i_u e > 0 \\ -e(t)/R & : \text{otherwise} \end{cases} \quad (60)$$

278 where

$$i_u = \mathbf{K}\mathbf{x} \quad (61)$$

279 And the generated average energy would be

$$\bar{P}_g^{\text{PG}} = \bar{P}_g^{\text{SA}} + R \mathcal{E} \{ (i_u + Y_c e)^2 - (i_u - i)^2 \} \quad (62)$$

280 which guarantees the inequality given by Eq. (57).

281 4. Numerical simulation

282 To verify the efficacy of the proposed two-body point absorber WEC, nu-
 283 merical simulation studies are carried out in this section. First, the param-
 284 eter values of the model used here are developed, then the power generation
 285 performance is assessed.

Table 1: Parameter values for numerical simulation

Parameter	Value	Parameter	Value
m_1	58,075 kg	m_2	34,515 kg
D_1	6.0 m	D_2	4.0 m
H_1	2.5 m	L	2.0 m
h	400 m	h_d	20 m
k_t	10,000 N/m	k_{PTO}	100,000 N/m
c_s	50 Ns/m	$C_{d,2}$	0.1
R	25 Ω	ρ	1,027 kg/m ³

286 4.1. Model development

287 The parameter values for the two-body point absorber used here is deter-
 288 mined based on the study conducted in [20], which are summarized in Table
 289 1.

290 The added mass and the radiation damping of the floating buoy and
 291 the submerged body calculated using WAMIT software [25] are shown in
 292 Figs. 4 and 5, respectively. The magnitude and the phase of the transfer
 293 functions given by Eq. (8) are shown in the figures as well. Moreover, the
 294 coupled added mass and radiation damping acting on the floating buoy from
 295 the submerged body, i.e., A_{12} and B_{12} and vice versa, i.e., A_{21} and B_{21} are
 296 shown in Fig. 6.

297 Then, from Eqs. (34) and (35), the frequency response data for $G_{a,1}$, $G_{z,1}$,
 298 $G_{a,2}$, and $G_{z,2}$ in Eqs. (36) and (37) are calculated as depicted in Figs. 7
 299 and 8 by solid lines. Then, to express these in state-space form as expressed
 300 by Eqs. (38) and (39), $G_{a,1}$ and $G_{a,2}$ are approximated with 5 zeros and 6
 301 poles, and $G_{z,1}$ with 2 zeros and 4 poles, and $G_{z,2}$ with 4 zeros and 5 poles.
 302 These numbers are chosen by trial and error so that the finite-dimensional
 303 models shown by dashed lines approximate the frequency response data very
 304 well. Note that the wave amplitude $a(t)$ is taken to be the wave amplitude
 305 $d = 10$ m ahead of the buoy in the propagation direction.

306 4.2. Inerter design

307 Next, the inertance m_s for the proposed WEC needs to be designed. In
 308 this study, this value is chosen so that the average power generation for the
 309 JONSWAP spectrum with $T_p = 6$ s is maximized when the SA controller is
 310 applied. While the damping coefficient of the drag force on the submerged

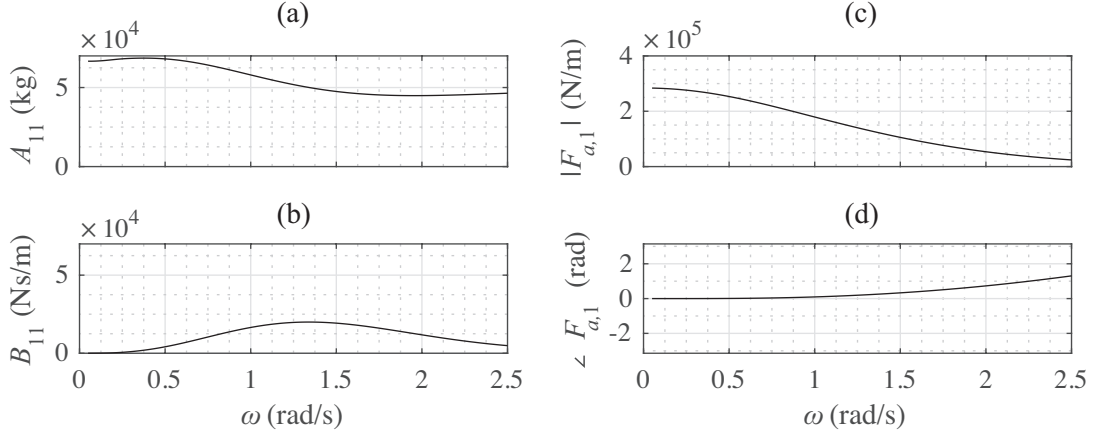


Figure 4: Hydrodynamic parameters for the heave mode of the floating buoy : (a) Added mass A_{11} , (b) Radiation damping B_{11} , (c) Magnitude of $F_{a,1}(\omega)$, (d) Phase of $F_{a,1}(\omega)$.

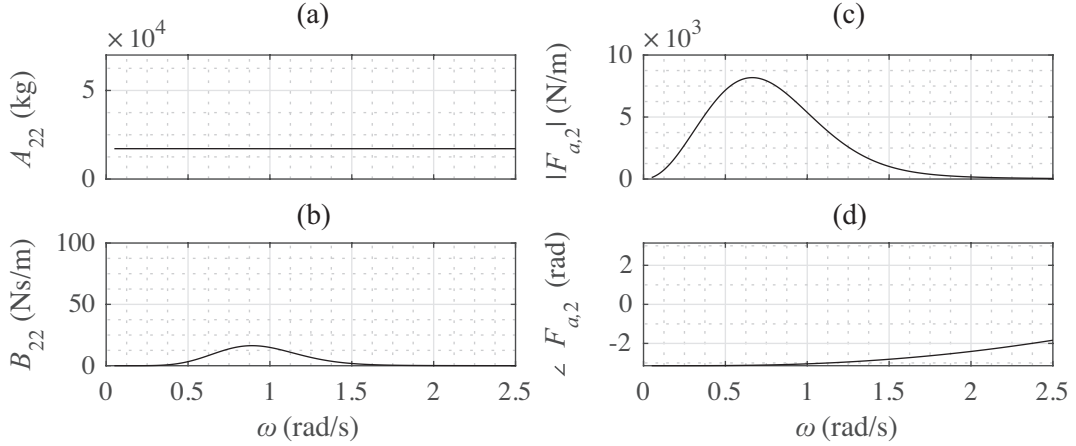


Figure 5: Hydrodynamic parameters for the heave mode of the submerged body : (a) Added mass A_{22} , (b) Radiation damping B_{22} , (c) Magnitude of $F_{a,2}(\omega)$, (d) Phase of $F_{a,2}(\omega)$.

body given by (16) depends on the standard deviation of the velocity $\dot{z}_2$. Thus, the value m_s is designed according to the flowchart as shown in Fig. 9 (a) here, in which $\sigma_{\dot{z}}$ and σ_z are standard deviations at the previous and current iterations, respectively. This flowchart is made based on the method to determine σ_z introduced in [26]. For this iteration process, the initial value for σ_z is set to $\sigma_{z0} = 0.1$ m/s and the range for m_s is set between $m_{s0} = 5,000$ kg and $m_{se} = 10,000$ kg and the best value is sought with respect to each

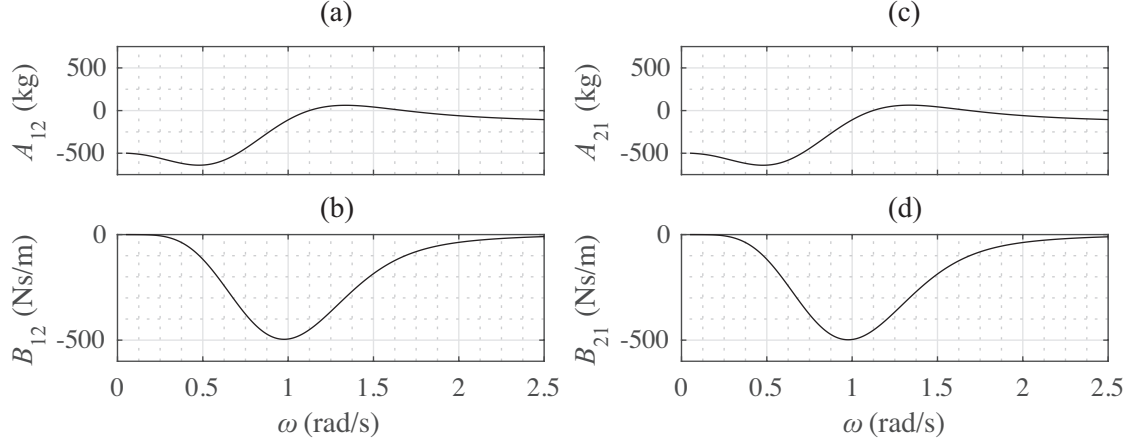


Figure 6: Coupled hydrodynamic parameters for the heave mode of the two bodies : (a) Added mass, A_{12} (b) Radiation damping B_{12} , (c) Added mass A_{21} , (d) Radiation damping B_{21} .

100 kg iteratively.

During the iteration, the standard deviation of $\dot{z}_2$ is calculated simply from

$$\sigma_{\dot{z}}^2 = \mathbf{G}^T \mathbf{S}_z \mathbf{G} \quad (63)$$

where $\mathbf{S}_z = \mathbf{S}_z^T > 0$ is the solution to the Lyapunov equation

$$(\mathbf{A} - Y_c \mathbf{B} \mathbf{C})^T \mathbf{S}_z + \mathbf{S}_z (\mathbf{A} - Y_c \mathbf{B} \mathbf{C}) + \mathbf{C}_z^T \mathbf{C}_z = \mathbf{0} \quad (64)$$

and

$$\dot{z}_2 = \mathbf{C}_z \mathbf{x} \quad (65)$$

Then, if the difference between the obtained $\sigma_{\dot{z}}$ and the previous value $\sigma_{\dot{z}}^-$ is larger than 0.001, the damping coefficient for the linearized drag force is recalculated using the newly obtained $\sigma_{\dot{z}}$ from Eq. (16). This procedure continues until the value of the standard deviation converges enough. And the average power for the SA control is calculated using the converged $\sigma_{\dot{z}}$ from Eq. (55)

The average power obtained in the process of the flowchart is plotted in Fig. 10. As can be seen, the generation performance peaks at $m_s = 6,900$ kg, which is used for the numerical simulation studies.

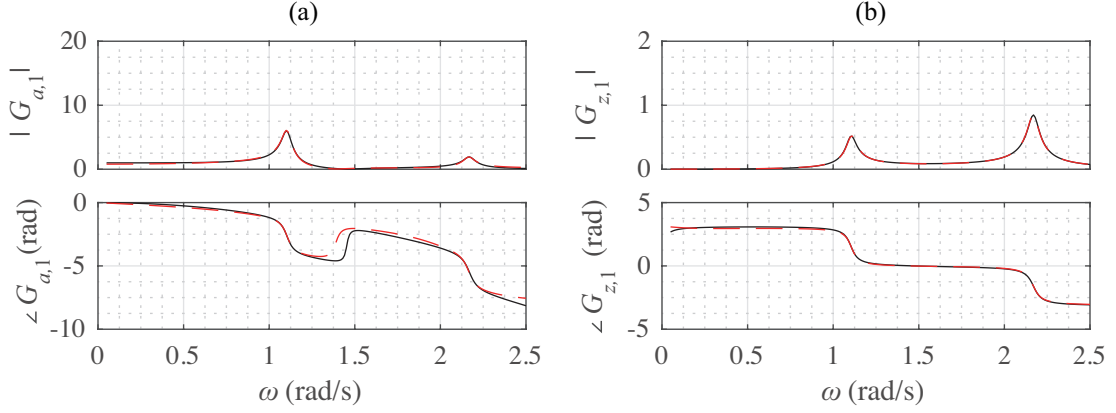


Figure 7: Frequency domain data (solid) and finite-dimensional approximation (dashed) :
(a) $G_{a,1}(\omega)$, (b) $G_{z,1}(\omega)$.

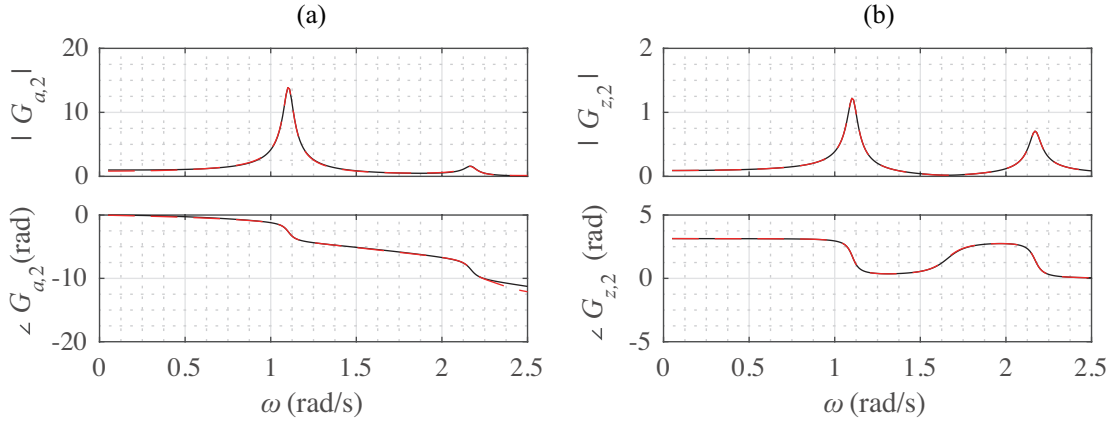


Figure 8: Frequency domain data (solid) and finite-dimensional approximation (dashed) :
(a) $G_{a,2}(\omega)$, (b) $G_{z,2}(\omega)$.

332 4.3. Drag force model

333 The damping coefficient $c_{v,2}$ for the linearized drag force acting on the
334 submerged body as expressed by Eq. (16) still needs to be obtained for
335 each peak wave period T_p . The inertance value is set to $m_s = 6,900$ kg as
336 determined before. Then we seek for the damping coefficient value for each
337 T_p of the JONSWAP spectrum based on the flowchart shown in Fig. 9 (b).
338 This is referred to the flowchart in [26] as well as Fig. 9 (a). For the process,
339 σ_z is first set to $\sigma_{z0} = 0.1$ m/s as well as before and $c_{v,2}$ is investigated in
340 the range of the peak wave period T_p from 2 s and 12 s, i.e., we set $T_{p0} = 2$

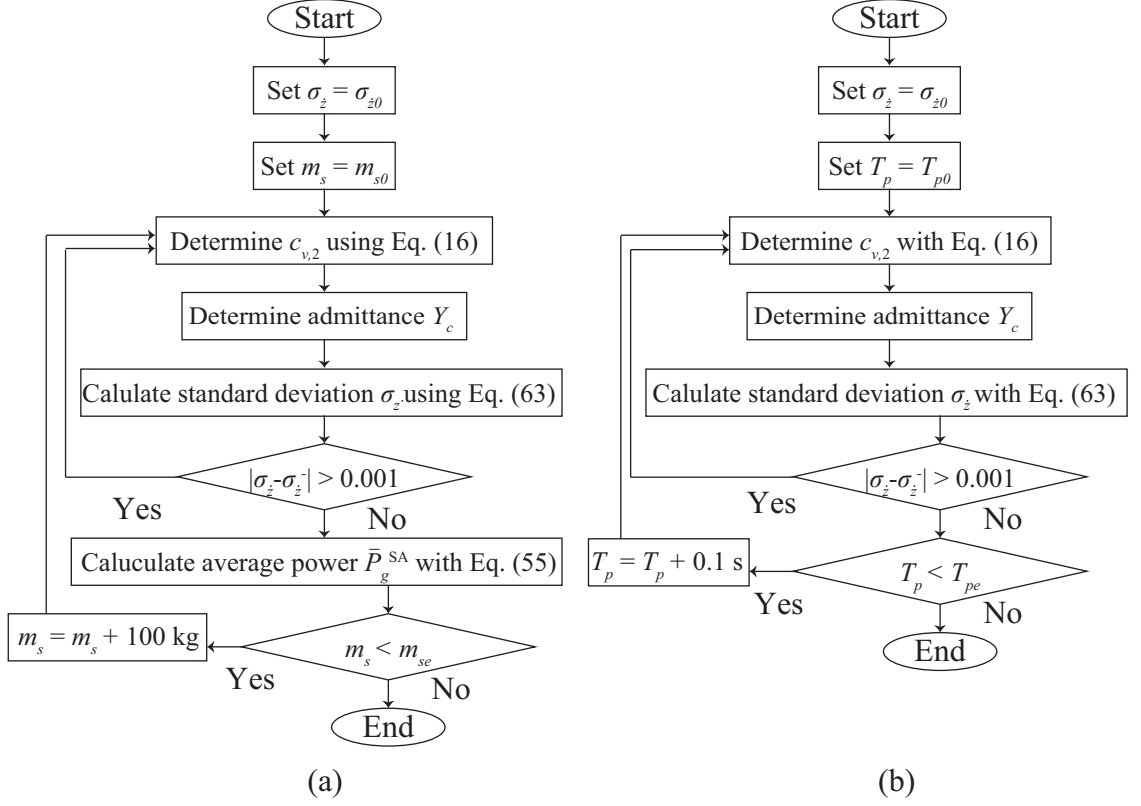


Figure 9: Flowchart: (a) Inerter design, (b) Drag force model.

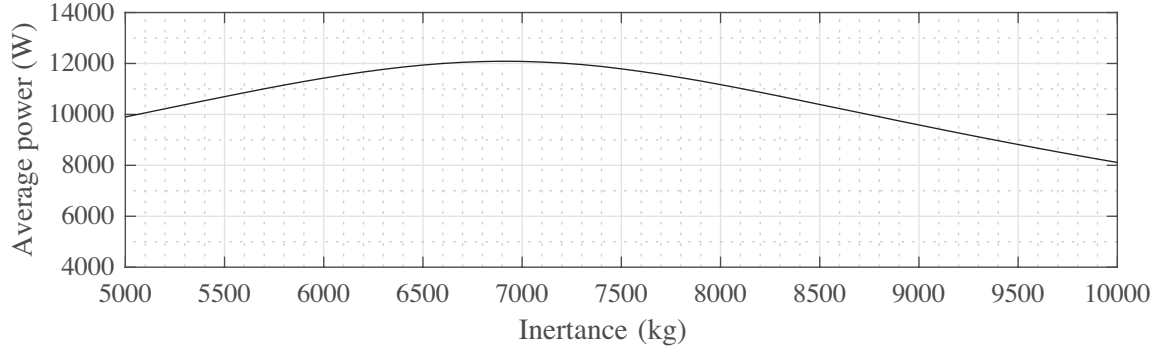


Figure 10: Average power vs inerter.

341 s and $T_{pe} = 12 \text{ s}$.

342 The result for the case of the two-body point absorber with a tuned inerter

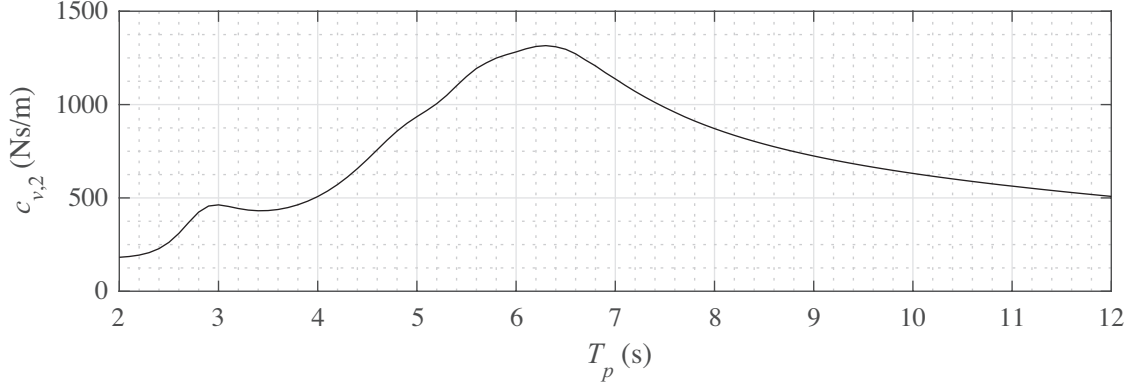


Figure 11: Damping coefficient $c_{v,2}$ for the drag force on the submerged body.

of $m_s = 6,900$ kg is shown in Fig. 11. We can find that $c_{v,2}$ is much larger than the radiation damping on the submerged body denoted by B_{22} and confirm that the drag force on the submerged body should not be ignored.

4.4. Results

Finally, the average power generations for the JONSWP spectrum with the peak wave period from 2 s to 12 s are calculated from the parameter values determined above. The results obtained from the SA and the PG controllers are compared in Fig. 12 (a), in which the proposed point absorber WEC with a tuned inerter and the conventional WEC cases are denoted by 2BwTI and 2B, respectively. The average power for the SA control cases can be given by the closed form expressed by Eq. (55), however, for consistency with the PG control cases, the average values shown in Fig. 12 (a) are calculated from Eq. (27) using the numerical simulation results for 100,000 s. The admittances for the SA controller are compared in Fig. 12 (b) which shows the difference of the optimized admittance between the proposed and the conventional WECs. For the assessment of the conventional WEC, the same parameters as the proposed WEC are employed except that $m_s = 50$ kg. As you can see in Fig. 12 (a), the proposed WEC shows the better power generation performance than the conventional WEC in the wide range of the peak wave period T_p even though the effectiveness deteriorates in a certain range of the lower period. Especially, the superiority of the proposed WEC is notable around $T_p = 6$ s to which the inerter of the proposed WEC is designed to be tuned. Also, the PG controller works well to improve the power generation especially for the proposed WEC.

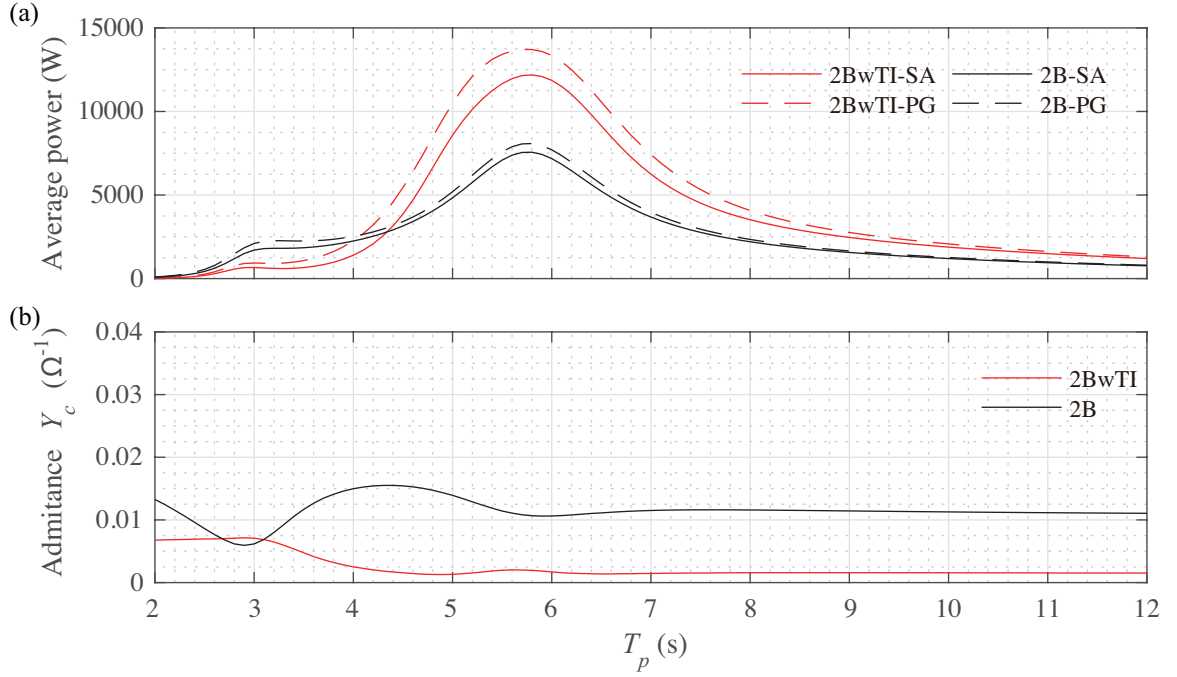


Figure 12: Results. (a) Power generation, (b) Admittance.

367 5. Conclusions

368 This article introduced the two-body point absorber WEC with a tuned
 369 inerter and developed the detailed analytical model including the coupled
 370 force between the two bodies. Then its effectiveness was verified by compar-
 371 ing the conventional two-body point absorber WEC through the numerical
 372 simulation studies. The results showed that the tuned inerter mechanism
 373 not only increased the power generation performance but also broadened the
 374 effective range of the wave period. Moreover, this research showed that the
 375 performance of the proposed WEC was improved further by controlling the
 376 current of the generator.

377 However, the optimum design including the shapes of the floating buoy
 378 and the submerged body should be examined more thoroughly, for example,
 379 by a CFD (computational fluid dynamics) technique. And various control
 380 algorithms such as reactive controllers should be applied to the generator
 381 current of the proposed WEC or more effective controllers need to be de-
 382 veloped. In addition, experimental verification is still required to show the
 383 reliability of the proposed WEC.

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